

## **CHAPTER (8)**

# **TIME VARYING ELECTRIC AND MAGNETIC FIELDS**

## Faraday's law:

Due to the first experiment of Faraday, we can say that a time-varying magnetic field produces an electromotive force (emf) which may establish a current in a suitable closed circuit. Faraday's law is stated as

$$emf = - \frac{\partial \psi_m(t)}{\partial t} \quad [V] \quad (1)$$

A non zero value of  $\left(\frac{\partial \psi_m(t)}{\partial t}\right)$  may result from any of the following situations:

- A time-changing flux linking a stationary closed path.
- Relative motion between a steady flux and a closed path.
- A combination of the two previous situations.

### Note:

1- The minus sign is an indication that the **emf** is in such a direction as to produce a current whose flux, if added to the original flux, would reduce the magnitude of the **emf**.

This statement that the induced voltage acts to produce an opposing flux is known as **Lenz's law**.

2- For  $N$  – turns filamentary conductor closed path

$$emf = - N \frac{\partial \psi_m(t)}{\partial t} \quad (2)$$

The emf is considered as the voltage about a specific closed path, and it is defined as

$$e.m.f = \oint \vec{E} \cdot d\vec{l} = - \frac{\partial \psi_m}{\partial t} \quad (3)$$

In electro-statics, equation (3) must lead to zero potential difference about a closed path. In time varying fields, the line integral leads to an emf or a potential difference.

Equation (3) can be written as

$$e.m.f = \oint \vec{E} \cdot d\vec{l} = - \frac{\partial \psi_m}{\partial t} = - \frac{\partial}{\partial t} \iint \vec{B} \cdot d\vec{s} \quad (4)$$

In general, Faraday's law manifests itself in either or both a stationary circuit linked by a time varying magnetic flux, such as a transformer, or the magnetic flux may be static, but the circuit is moving relative to the flux in such a way as to produce a time varying flux enclosed by the circuit. A rotating machine generates an emf by the latter mechanism.

### 1<sup>st</sup> Maxwell's equation

Applying Stock's theorem on equation (4)

$$e.m.f = \oint \vec{E} \cdot d\vec{l} = \iint \nabla \times \vec{E} \cdot d\vec{s} = - \frac{\partial}{\partial t} \iint \vec{B} \cdot d\vec{s}$$

So, **1<sup>st</sup> Maxwell's equation in differential form is**

$$\nabla \times \vec{E} = - \frac{\partial \vec{B}}{\partial t} \quad (5)$$

And, 1<sup>st</sup> Maxwell's equation in integral form is

$$\oint \vec{E} \cdot d\vec{l} = \iint \nabla \times \vec{E} \cdot d\vec{s} = - \frac{\partial}{\partial t} \iint \vec{B} \cdot d\vec{s} \quad (6)$$

### Displacement current density $\vec{J}_d$

It is shown from 1<sup>st</sup> Maxwell's equation in differential form (5)

$$\nabla \times \vec{E} = - \frac{\partial \vec{B}}{\partial t}$$

that a time varying magnetic field produces an electric field  $\vec{E}$ . From the definition of the curl, the electric field  $\vec{E}$  has the special property of circulation; its line integral about a general closed path is not zero.

Now we will consider the time varying electric field. From Curl law as it applies to steady magnetic fields, where

$$\nabla \times \vec{H} = \vec{J}_c \quad (7)$$

Taking the divergence of both sides of equation (7),

$$\nabla \cdot \nabla \times \vec{H} = \nabla \cdot \vec{J}_c \quad (8)$$

Mathematically, the left hand side of equation (8) equals zero ( $\nabla \cdot \nabla \times \vec{H} = 0$ ), while the right hand side of equation (8), from continuity equation (conservation of charge) equals to

$$\nabla \cdot \vec{J}_c = - \frac{\partial \rho_v}{\partial t} \quad (9)$$

So, equation (8) becomes

$$\nabla \cdot \nabla \times \vec{H} = \nabla \cdot \vec{J}_c = -\frac{\partial \rho_v}{\partial t} \quad (10)$$

Since in equation (10), the L.H.S = 0, but the R.H.S  $\neq 0$ , so there is a missing term that appear in time varying field as that appear in first Maxwell's equation  $\left(-\frac{\partial \vec{B}}{\partial t}\right)$ , in order to fulfill that R.H.S = L.H.S. This missing term that must be added to the L.H.S is  $\left(+\frac{\partial \rho_v}{\partial t}\right)$ . So

$$\nabla \cdot \nabla \times \vec{H} = -\frac{\partial \rho_v}{\partial t} + \frac{\partial \rho_v}{\partial t} \quad (11)$$

Using the divergence law,  $\nabla \cdot \vec{D} = \rho_v$ , equation (11) becomes

$$\nabla \cdot \nabla \times \vec{H} = -\frac{\partial \rho_v}{\partial t} + \frac{\partial}{\partial t} (\nabla \cdot \vec{D})$$

$$\nabla \cdot \left[ \nabla \times \vec{H} - \frac{\partial \vec{D}}{\partial t} \right] = -\frac{\partial \rho_v}{\partial t} = \nabla \cdot \vec{J}_c$$

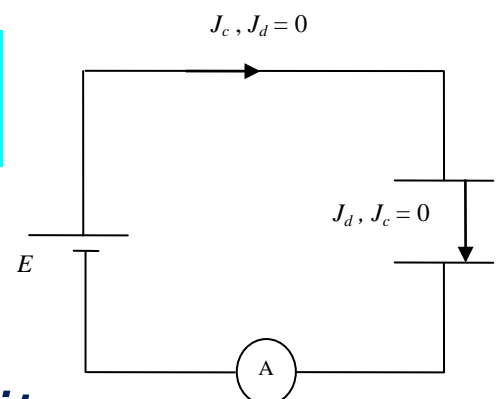
So,

$$\nabla \times \vec{H} = \vec{J}_c + \frac{\partial \vec{D}}{\partial t}$$

$$\nabla \times \vec{H} = \vec{J}_c + \vec{J}_d$$

$\vec{J}_c$  is the conduction current density

$\vec{J}_d$  is the displacement current density



So, **2<sup>nd</sup> Maxwell's equation in differential form is**

$$\nabla \times \vec{H} = \vec{J}_c + \vec{J}_d \quad (12)$$

And, **2<sup>nd</sup> Maxwell's equation in integral form is**

$$\oint \vec{H} \cdot d\vec{l} = \iint \nabla \times \vec{H} \cdot d\vec{s} = \iint \frac{\partial \vec{D}}{\partial t} \cdot d\vec{s} + \iint \vec{J}_c \cdot d\vec{s} \quad (13)$$

**3<sup>rd</sup> Maxwell's equation:**

$\oint \vec{B} \cdot d\vec{S} = 0 \rightarrow$  3<sup>rd</sup> Maxwell's equation in Integral form

**Applying the divergence theorem**

$$\oint \vec{B} \cdot d\vec{s} = \iiint \nabla \cdot \vec{B} dv = 0$$

So, **3<sup>rd</sup> Maxwell's equation in differential form is**

$$\nabla \cdot \vec{B} = 0 \quad (14)$$

And, **3<sup>rd</sup> Maxwell's equation in integral form is**

$$\oint \vec{B} \cdot d\vec{s} = \iiint \nabla \cdot \vec{B} dv = 0 \quad (15)$$

## 4<sup>th</sup> Maxwell's equation:

Applying the divergence theorem on the electric flux density  $\vec{D}$ , we have

$$\oint \vec{D} \cdot \vec{ds} = Q_{en} = \iiint \nabla \cdot \vec{D} \, dv = \iiint \rho_v \, dv$$

So, 4<sup>th</sup> Maxwell's equation in differential form is

$$\nabla \cdot \vec{D} = \rho_v \quad (16)$$

And, 4<sup>th</sup> Maxwell's equation in integral form is

$$\oint \vec{D} \cdot \vec{ds} = Q_{en} = \iiint \nabla \cdot \vec{D} \, dv = \iiint \rho_v \, dv \quad (17)$$

## Summary:

| Maxwell's equation   |   |
|--|---|
| Differential form  | Integral form   |
| $\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$ | $\oint \vec{E} \cdot \vec{dl} = \iint \nabla \times \vec{E} \cdot \vec{ds} = -\frac{\partial}{\partial t} \iint \vec{B} \cdot \vec{ds}$                                 |
| $\nabla \times \vec{H} = \vec{J}_c + \vec{J}_d$                | $\oint \vec{H} \cdot \vec{dl} = \iint \nabla \times \vec{H} \cdot \vec{ds} = \iint \frac{\partial \vec{D}}{\partial t} \cdot \vec{ds} + \iint \vec{J}_c \cdot \vec{ds}$ |
| $\nabla \cdot \vec{D} = \rho_v$                                | $\oint \vec{D} \cdot \vec{ds} = Q_{en} = \iiint \nabla \cdot \vec{D} \, dv = \iiint \rho_v \, dv$   |
| $\nabla \cdot \vec{B} = 0$                                     | $\oint \vec{B} \cdot \vec{ds} = \iiint \nabla \cdot \vec{B} \, dv = 0$  |

$$\vec{J}_d = \frac{\partial \bar{D}}{\partial t} = \frac{\epsilon_0 \partial \bar{E}}{\partial t}$$

$$\nabla \cdot \bar{J}_c + \frac{\partial \rho_v}{\partial t} = 0$$

$$I_d = \vec{J}_d \cdot \vec{A}$$



### Example:

A coaxial cable of  $\epsilon_r = 4$ , the inner conductor radius = 1mm and the outer conductor radius = 5mm. Find the displacement current between the two conductors per meter if the voltage difference equal  $100 \cos(12\pi \cdot 10^6 t)$  volt.

### Solution:

#### 1- Using circuit concept

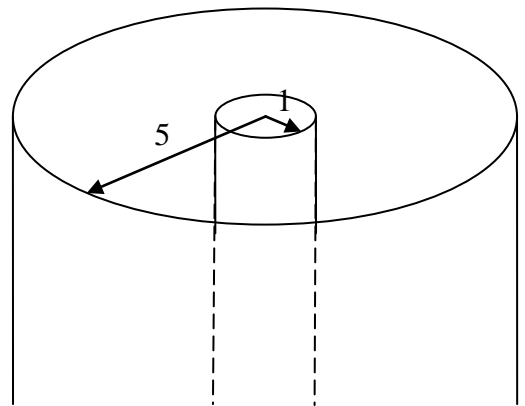
For a coaxial cable:

$$C = \frac{2\pi\epsilon}{\ln\left(\frac{b}{a}\right)} = \frac{2\pi * 8.85 * 10^{-12} * 4}{\ln\left(\frac{5}{1}\right)}$$

$$I_d = C \frac{dV}{dt}$$

$$\frac{dV}{dt} = 100 \sin(12\pi * 10^6 t) * 12\pi * 10^6$$

$$I_d = \frac{2\pi * 8.85 * 10^{-12} * 4}{\ln\left(\frac{5}{1}\right)} 100 \sin(12\pi * 10^6 t) * 12\pi * 10^6 \text{ A/m}$$



#### 2- Using electro-magnetic field concept

The displacement current density  $\vec{J}_d$  can be determined from  $\vec{J}_d = \frac{\partial \vec{D}}{\partial t}$ , after determining  $\vec{D}$  by applying the Gauss's law as follows:

1)  $\oint \vec{D} \cdot d\vec{s} = Q_{en}$

2) choice of Gaussian surface

3)  $Q_{en} = \rho_l l$

4)  $\oint \vec{D} \cdot d\vec{s} = 2\pi r_c l D$

5)  $2\pi r_c l D = \rho_l l$

6)  $\vec{D} = \hat{a}_{r_c} \frac{\rho_l}{2\pi r_c} \text{ C/m}^2$

7)  $\vec{E} = \hat{a}_{r_c} \frac{\rho_l}{2\pi\epsilon_0\epsilon_r r_c} \text{ N/C}$

In order to determine  $\rho_l$  in terms of the potential difference  $V$ , the potential  $V$  is determined as

$$V_{ab} = - \int_b^a \vec{E} \cdot \vec{dl}$$

$$V_{ab} = - \int_5^1 \hat{a}_{r_c} \frac{\rho_l}{2\pi\epsilon_0\epsilon_r r_c} \cdot \hat{a}_{r_c} dr_c$$

$$100 \cos (12\pi \cdot 10^6 t) = \frac{\rho_l}{2\pi\epsilon_0\epsilon_r} \ln (5)$$

$$\rho_l = \frac{200 \pi\epsilon_0\epsilon_r \cos (12\pi \cdot 10^6 t)}{\ln (5)}$$

So,

$$\vec{D} = \hat{a}_{r_c} \frac{\rho_l}{2\pi r_c} = \hat{a}_{r_c} \frac{200 \pi\epsilon_0\epsilon_r \cos (12\pi \cdot 10^6 t)}{2\pi r_c \ln (5)}$$

$$\vec{J}_d = \frac{\partial \vec{D}}{\partial t} = - \hat{a}_{r_c} \frac{200 \pi\epsilon_0\epsilon_r (12\pi \cdot 10^6) \sin (12\pi \cdot 10^6 t)}{2\pi r_c \ln (5)}$$

$$I_d = \iint \vec{J}_d \cdot \vec{ds}$$

$$I_d = \int_0^{2\pi} \int_0^l - \hat{a}_{r_c} \frac{200 \pi\epsilon_0\epsilon_r (12\pi \cdot 10^6) \sin (12\pi \cdot 10^6 t)}{2\pi r_c \ln (5)} \cdot \hat{a}_{r_c} r_c d\phi dz$$

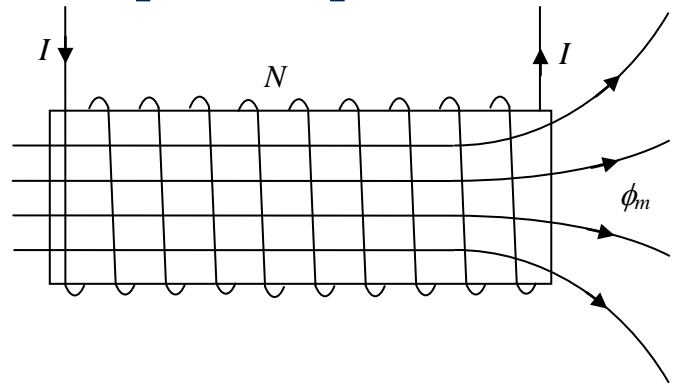
$$I_d = \frac{200 \pi\epsilon_0\epsilon_r (12\pi \cdot 10^6) \sin (12\pi \cdot 10^6 t)}{\ln (5)} l$$

$$\frac{I_d}{l} = \frac{200 \pi\epsilon_0\epsilon_r (12\pi \cdot 10^6) \sin (12\pi \cdot 10^6 t)}{\ln (5)} \text{ A/m}$$

## Inductance:

When dealing with time – varying current and thus time – varying magnetic fields, the inductance of the inductor as a circuit concept becomes quite important and is defined as

$$L = \frac{\Lambda}{I} \quad [H]$$



Where:

$\Lambda$  (lambda) is the total flux linkage of the inductors and  $I$  is the current flowing in the inductor.

A flux linkage of one exists when 1 Wb of flux links one turn of the inductor.

So, if all the flux link all the turns, the total flux linkage is equal to

$$\Lambda = N \psi_m \quad [\text{Wb turns}]$$

Where,  $\psi_m$  is the total flux produced by the inductor

$L$  is the self inductance

If the coil has  $N$  turns:

$$L = \frac{\Lambda}{I} = \frac{N \psi_m}{I}$$

### Example:

Find the self inductance of the solenoid shown in figure.

### Solution:

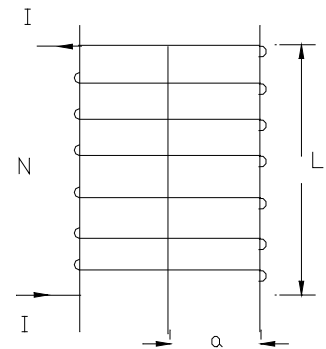
$$L = \frac{\Lambda}{I} = \frac{N \Psi_m}{I}$$

$$\Psi_m = \iint \vec{B} \cdot d\vec{s} = B A = B (\pi a^2)$$

$$\vec{B} = \mu_0 \mu_r \vec{H}$$

The magnetic field intensity  $\vec{H}$  can be determined by applying Ampere's circuital law as follows:

- (1)  $\oint \vec{H} \cdot d\vec{l} = I_{en}$
- (2) *Choice of Amperian loop*
- (3)  $I_{en} = N I$
- (4)  $\oint \vec{H} \cdot d\vec{l} = H l$
- (5)  $\vec{H} = \frac{N I}{l} \hat{a}_z$



So,

$$\vec{B} = \mu_0 \mu_r \vec{H} = \mu_0 \mu_r \frac{N I}{l} \hat{a}_z$$

$$\Psi_m = B (\pi a^2) = \mu_0 \mu_r \frac{N I}{l} (\pi a^2)$$

$$L = \frac{\Lambda}{I} = \frac{N \Psi_m}{I} = \mu_0 \mu_r \frac{N^2}{l} (\pi a^2) \quad [\text{H}]$$

### Example:

Find the self inductance of the toroid shown.

### Solution:

$$L = \frac{\Lambda}{I} = \frac{N \psi_m}{I}$$

The magnetic field intensity  $\vec{H}$  can be determined by applying Ampere's circuital law as follows:

- (1)  $\oint \vec{H} \cdot d\vec{l} = I_{en}$
- (2) *Choice of Amperian loop*
- (3)  $I_{en} = N I$
- (4)  $\oint \vec{H} \cdot d\vec{l} = H 2\pi r_c$
- (5)  $\vec{H} = \frac{N I}{2\pi r_c} \hat{a}_\phi$

So,

$$\vec{B} = \mu_0 \mu_r \vec{H} = \mu_0 \mu_r \frac{N I}{2\pi r_c} \hat{a}_\phi$$

$$\text{Average path radius } r_{c|average} = b = \frac{a+c}{2}$$

$$\vec{B}_{av} = \psi_m = \hat{a}_\phi$$

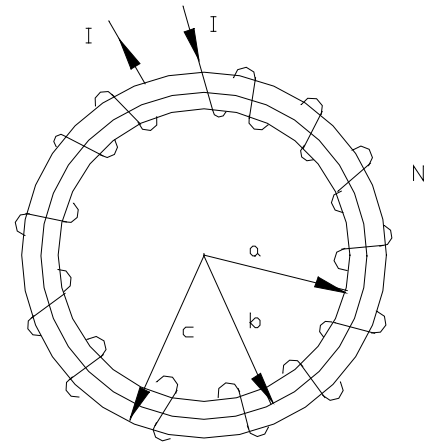
$$\psi_m = B_{av} A$$

$$A = \pi \left( \frac{c-a}{2} \right)^2$$

$$\psi_m = \mu_0 \mu_r \frac{N I}{2\pi b} \pi \left( \frac{c-a}{2} \right)^2$$

$$\psi_m = \mu_0 \mu_r \frac{N I (c-a)^2}{4(c+a)}$$

$$L = \frac{\Lambda}{I} = \frac{N \psi_m}{I} = \mu_0 \mu_r \frac{N^2 (c-a)^2}{4(c+a)} \quad \text{Henry}$$

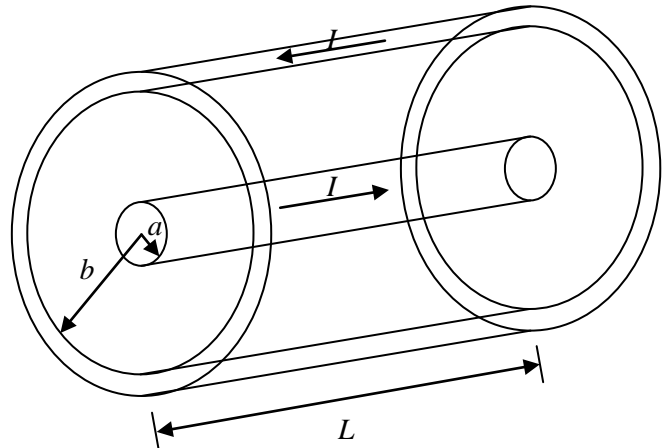


### Example:

Find the self inductance per unit length of a T.L. (coaxial cable).

### Solution:

$$L = \frac{\Lambda}{I} = \frac{N \Psi_m}{I} \quad N = 1$$



The magnetic field intensity  $\vec{H}$  can be determined by applying Ampere's circuital law as follows:

- (1)  $\oint \vec{H} \cdot d\vec{l} = I_{en}$
- (2) *Choice of Amperian loop*
- (3)  $I_{en} = I$
- (4)  $\oint \vec{H} \cdot d\vec{l} = H 2\pi r_c$
- (5)  $\vec{H} = \frac{I}{2\pi r_c} \hat{a}_\phi$

So,

$$\vec{B} = \mu_0 \mu_r \vec{H} = \mu_0 \mu_r \frac{I}{2\pi r_c} \hat{a}_\phi$$

$$\Psi_m = \iint \vec{B} \cdot d\vec{s} = \int_0^l \int_a^b \mu_0 \mu_r \frac{I}{2\pi r_c} \hat{a}_\phi \cdot \hat{a}_\phi dr_c dz$$

$$\Psi_m = \frac{\mu_0 \mu_r I l}{2\pi} \ln(r_c)_a^b = \frac{\mu_0 \mu_r I l}{2\pi} \ln\left(\frac{b}{a}\right)$$

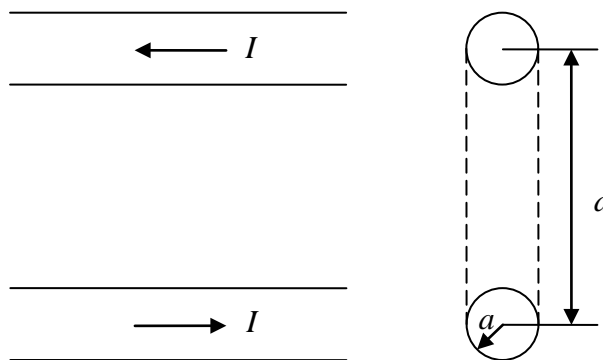
$$L = \frac{\Lambda}{I} = \frac{\Psi_m}{I} = \frac{\mu_0 \mu_r l}{2\pi} \ln\left(\frac{b}{a}\right) \quad [\text{H}]$$

$$\frac{L}{l} = \frac{\mu_0 \mu_r}{2\pi} \ln\left(\frac{b}{a}\right) \quad [\text{H/m}]$$

**Example:**

Find the inductance per meter for two parallel wires.

**Solution:**



$$L = \frac{\Lambda}{I} = \frac{N \Psi_m}{I} \quad N = 1$$

The magnetic field intensity  $\vec{H}$  can be determined by applying Ampere's circuital law as follows:

- (1)  $\oint \vec{H} \cdot d\vec{l} = I_{en}$
- (2) *Choice of Amperian loop*
- (3)  $I_{en} = I$
- (4)  $\oint \vec{H} \cdot d\vec{l} = H 2\pi r_c$
- (5)  $\vec{H} = \frac{I}{2\pi r_c} \hat{a}_\phi$

So,

$$\vec{B} = \mu_0 \mu_r \vec{H} = \mu_0 \mu_r \frac{I}{2\pi r_c} \hat{a}_\phi$$

$$\Psi_m = \iint \vec{B} \cdot d\vec{s} = \int_0^l \int_a^{d-2a} \mu_0 \mu_r \frac{I}{2\pi r_c} \hat{a}_\phi \cdot \hat{a}_\phi dr_c dz$$

$$\Psi_m = \frac{\mu_0 \mu_r I l}{2\pi} \ln(r_c) \frac{d-2a}{a} = \frac{\mu_0 \mu_r I l}{2\pi} \ln\left(\frac{d-2a}{a}\right)$$

$$L = \frac{\Lambda}{I} = \frac{\Psi_m}{I} = \frac{\mu_0 \mu_r l}{2\pi} \ln\left(\frac{d-2a}{a}\right) \quad [\text{H}]$$

### Mutual inductance:

Mutual inductance exists between two magnetic circuits that share a common flux linkage. So, the mutual inductance is due to the influence of one circuit on another and vice versa.

$$M_{12} = \frac{\Lambda_{12}}{I_1} = \frac{N_2 \Psi_{m12}}{I_1}$$

Where,  $\Lambda_{12}$  is the linkage of circuit 2 produced by  $I_1$  in circuit 1.

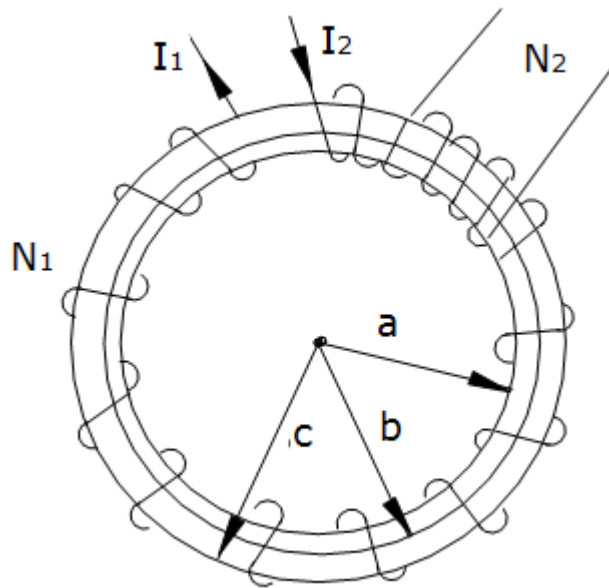
For a linear magnetic medium, it can be shown that  $M_{12} = M_{21}$ .

$$M_{21} = \frac{\Lambda_{21}}{I_2} = \frac{N_1 \Psi_{m21}}{I_2}$$



**Example:**

Find  $M_{12}$ .



**Solution:**

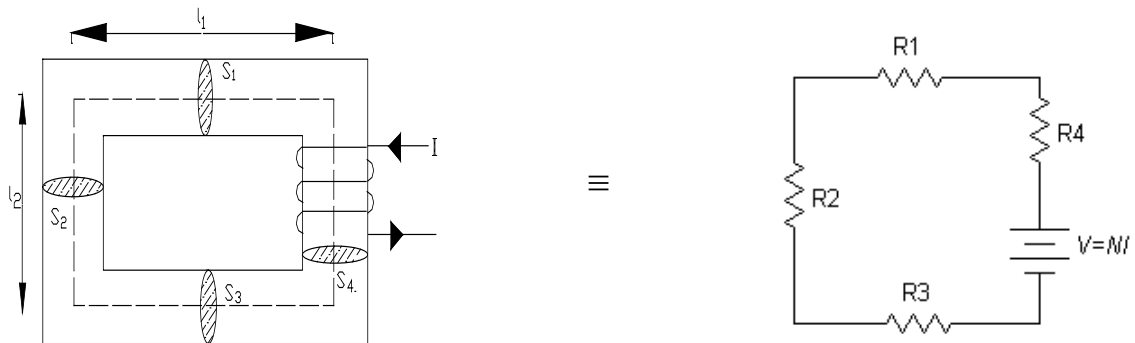
$$\psi_{12} = \frac{\mu_0 \mu_r N_1 I_1 (c - a)^2}{4(c + a)}$$

$$M_{12} = \frac{N_2 \psi_{12}}{I_1}$$

$$\therefore M_{12} = \frac{\mu_0 \mu_r N_1 N_2 I_1 (c - a)^2}{4 I_1 (c + a)} = \frac{\mu_0 \mu_r N_1 N_2 (c - a)^2}{4(c + a)}$$

## Magnetic Circuits:

### Example:



$$NI = R\psi_m$$

**R = Reluctance**

= The impedance of magnetic material for flow of magnetic flux

$$R = \frac{l}{\mu_0 \mu_r S} \rightarrow l = \text{magnetic path length}, S = \text{cross sectional area}$$

$$R_1 = \frac{l_1}{\mu_0 \mu_r S_1},$$

$$R_2 = \frac{l_2}{\mu_0 \mu_r S_2},$$

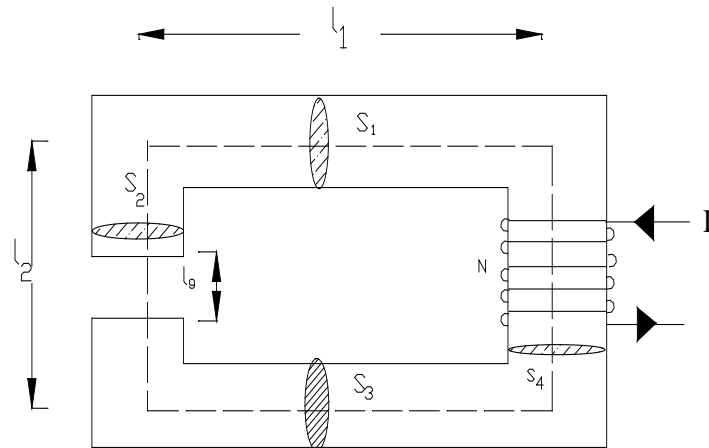
$$R_3 = \frac{l_3}{\mu_0 \mu_r S_3},$$

$$R_4 = \frac{l_4}{\mu_0 \mu_r S_4}$$

$$\therefore \psi_m (R_1 + R_2 + R_3 + R_4) = NI$$

$$\therefore \psi_m = \frac{NI}{R_1 + R_2 + R_3 + R_4} = B \cdot A = B \cdot S$$

**Example:**

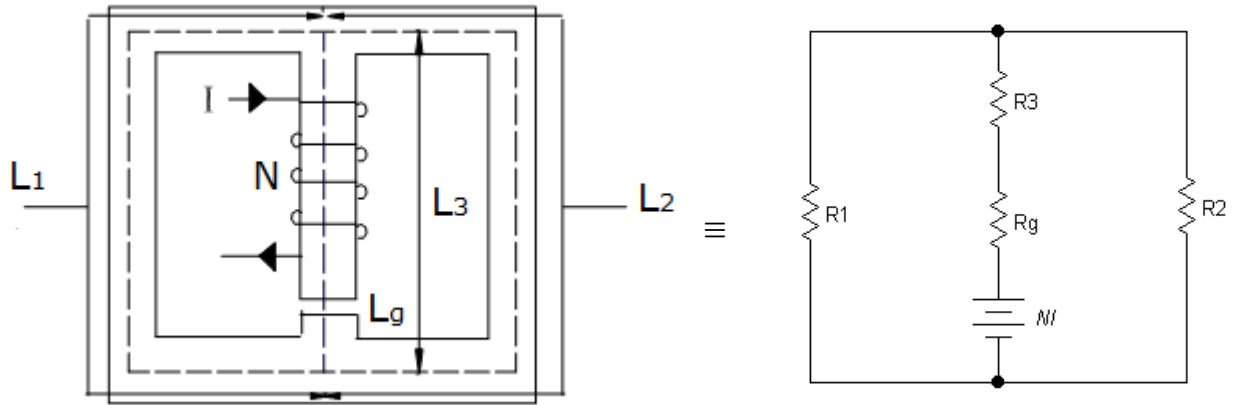


$$R_g = \frac{l_g}{\mu_0 S_2}$$

$$\therefore \psi_m (R_1 + R_2 + R_3 + R_4 + R_g) = NI$$

$$\psi_m = \frac{NI}{(R_1 + R_2 + R_3 + R_4 + R_g)} = B \cdot S$$

**Example:**



$$\Psi_{m3} = \Psi_{m1} + \Psi_{m2}$$

$$R_t = R_g + R_3 + \frac{R_1 R_2}{R_1 + R_2}$$

$$\Psi_{m3} \cdot R_t = NI \quad \rightarrow \quad \Psi_{m3} = \frac{NI}{R_t}$$

$$\therefore \Psi_{m1} = \Psi_{m3} \cdot \frac{R_2}{R_1 + R_2} \quad ,$$

$$\Psi_{m2} = \Psi_{m3} \cdot \frac{R_1}{R_1 + R_2}$$